

DETAILED POL SYNTH PROGRAMMING GUIDE

Syd Kreitzman

----- rev history -----

Rev Nov 20/2003 ... original spec

Rev Dec 1 /2003 ... added functions idle_freq idle_iq
added detailed procedure to calculate the frequency sweep values

Rev Feb 26/2004 ... added gate control logic into register 2C
and moved old 2C functions to 2D, old 2D function to 2E (see * lines)

Rev May 10/2004 ... added section on how to do the complex modulation; srk

Rev July 19/2004 ... corrected and updated complex modulation section; srk

Rev Sept 10/2004 ... added functionality for iq data recycling by introducing five new
Registers Ncmx, Nc1f, Nc3f, Nc5f, Ncfref into locations 30-31
Also introduced the modulation function parameters A and α to define
the modulation function properties (i.e. linewidth and ending amplitude); srk

Rev Sept 30/2004 ... corrected for the fact that the Niq data gets clocked in at $4 \cdot N_{cic} \cdot 25 \cdot 10^{(-9)}$
seconds (i.e. every $4 \cdot N_{cic}$ clock periods) ... thereby correcting calculations
for pulse lengths, bandwidths etc; srk

Rev Feb 28/2005 ... added control of output scale factorto give amplitude control to modulated
functions without having to reprogram the entire I/Q array. A new global
control function "scale" has been added. The scaled value is $2^{7 \cdot \text{scale}}$,
where scale is an eight bit value from 0 -> 255. The default value of scale
is $(256/\sqrt{2})=181$.

Rev April 25/2005 ... Added Hermite modulation function.

----- end of rev history -----

Local Control Register Level Functionality

Function	Value	Register	Access
1f-qm-on	0x	01	r/w
1f-qm-off	1x	01	r/w
1f-cic	2-ffx	02	r/w
1f-scale	0x-ffx	03	r/w
1f-n iq	0x-7ffx	04-05	r/w
1f-iq_ptr		06-07	r only
1f-on	1,1b	2C bits 0,1	r/w
1f-off	0,0b	2C bits 0,1	r/w
1f-gate-t	1,0b	2C bits 0,1	r/w
1f-gate-f	0,1b	2C bits 0,1	r/w
Nc1f	1x-01x	31 bits 0-5	r/w 0x is not allowed
3f-qm-on	0x	09	r/w
3f-qm-off	1x	09	r/w
3f-cic	2-ffx	0A	r/w
3f-scale	0x-ffx	0B	r/w

3f-n_iq	0x-7ffx	0C-0D	r/w
3f-iq_ptr		0E-0F	r only
3f-on	1,1b	2C bits 2,3	r/w
3f-off	0,0b	2C bits 2,3	r/w
3f-gate-t	1,0b	2C bits 2,3	r/w
3f-gate-f	0,1b	2C bits 2,3	r/w
Nc3f	1x-01x	32 bits 0-5	r/w 0x is not allowed

5f-qm-on	0x	11	r/w
5f-qm-off	1x	11	r/w
5f-cic	2-ffx	12	r/w
5f-scale	0x-ffx	13	r/w
5f-n_iq	0x-7ffx	14-15	r/w
5f-iq_ptr		16-17	r only
5f-on	1,1b	2C bits 4,5	r/w
5f-off	0,0b	2C bits 4,5	r/w
5f-gate-t	1,0b	2C bits 4,5	r/w
5f-gate-f	0,1b	2C bits 4,5	r/w
Nc5f	1x-01x	33 bits 0-5	r/w 0x is not allowed

fref-qm-on	0x	19	r/w
fref-qm-off	1x	19	r/w
fref-cic	2-ffx	1A	r/w
fref-scale	0x-ffx	1B	r/w
fref-n_iq	0x-7ffx	1C-1D	r/w
fref-iq_ptr		1E-1F	r only
fref-on	1,1b	2C bits 6,7	r/w
fref-off	0,0b	2C bits 6,7	r/w
fref-gate-t	1,0b	2C bits 6,7	r/w
fref-gate-f	0,1b	2C bits 6,7	r/w
Ncfref	1x-01x	34 bits 0-5	r/w 0x is not allowed
fref-freqx	0-fffffffx	20-23	r/w write order: 20 first 23 last
fref-freqx = (fref_Hz*2^32/(4x10^7))x=(fref_Hz*(107.3741824))x=(fref_Hz/finc)x			
finc = .009313226Hz			
fmax = 4*10^7 - finc			

n_fsweep	0-3ff	24-25	r/w
fsweep_ptr		26-27	r only

iq-end_idle	?fx	28 bits 1-4only	r/w
iq-end_niq	?0x	28 bits 1-4 only	r/w
fsweep-end_idle	1?x	28 bit 5 only	r/w
fsweep-end_nfsweep	0?x	28 bit 5 only	r/w

or reg_28x = (iq-end*0fx)+(sweep-end*10x),
 where iq-end is 1 (iq-end_idle) or 0 (iq-end_niq)
 and sweep-end is 1 (sweep-end_idle) or 0 (sweep-end_nfsweep).
 reg_28x is calculated and written whenever the values of iq-end or sweep-end are changed.

RF_power_trip_thr	0-ffx	29	r/w
fsweep_int_strobe	0		w, r is always 0
fsweep_ptr_reset	0	2b	w, r is always 0
RF_power_trip_stat	0	2d	w cycle will reset, read is 1 or 0
VME_reset	0	2e	w cycle resets all vms and DDS registers to default
Ncmx	1x	30	max number of cycles available to cycle the iq pairs

Global Control Functions

qm-on > 1f-qm-on,3f-qm-on,5f-qm-on,frfef-qm-on

qm-off > 1f-qm-off,3f-qm-off,5f-qm-off,frfef-qm-off
n_iq 0x-7ffx > 1f-n_iq,3f-n_iq,5f-n_iq,frfef-n_iq all set to same value, default value is 2048 = 8FFx
iq_ptr displays all 1r,3f,5f,frfef-iq_ptr
cic_ir 2-ffx > 1f-cic, 3f-cic, 5f-cic, frfef-cic all set to same value
idle_freq (specify and load the value of idle_freq into 8FFC-F)
idle_iq (specify and load the i and q modulation default values into 1/3/5/7FFC-F)
Nc 1x-01x > Nc1f, Nc3f, Nc5f, Ncfref all set to same value Nc
scale 0x-ffx (i.e. 8bit 0-255), change 1f-scale, 3f-scale, 5-scale, frfef-scale to same scale value

Freq Sweep Loading

fsweep 0-ffffffx 8000-8FFC r/w n_fsweep locations to to loaded
 8FFD-8FFF r/w + the idle frequency

It is probably best to divide the freq steps into multiples of finc, so determine the closest start freq, the closest value of the requested freq step in units of finc (call this delta_fincx) and the number of dealta_fincs required to get to withing 1/2 finc of the requested ending frequency. Then you can program the first location with the closest initial start frequency, and a delta_fincx to that value successively for each 32bit frequency word. This means that the actual start, stop, and step frequencies are computed as well as the number (i.e. n_fsweep) steps to get there. These number will be slightly different than what the user requested ... but the frequency increments will be absolutely constant. The calculation of the frequency is the same as that indicated for frf_freq. See detailed instructions below.

IQ data 0000-1FFF	}	
2000-3FFF	}	All sets of registers are loaded with same set of data.
4000-5FFF	}	The 1/3/5/7FFE - 1/3/5/7FFF locations are always loaded
6000-7FFF	}	with the idle I/Q data pair.

Detailed notes to for frequency sweep programing:

- i) Determine the actual_start_frequency, it will be the closest frequency to the requested start_freq
- ii) Determine the actual_frequency_step, it will the frequency closest to the requested freq_step
- iii) Determinfine the actual_number_freq_steps to get to the first frequency beyond the requested stop_freq
- iv) Compute the actual_stop_frequency

actual_start_frequencyx = (int(start_frequency/finc))x
actual_start_frequency_hz = (int(start_frequency/finc))*finc

actual_frequency_stepx = (int(freq_step)/finc))x
actual_frequency_step_hz = (int(frq_step)/finc))*finc

actual_number_freq_steps = int((stop_freq - start_freq)/actual_frequency_step_hz)
actual_stop_freq_hz = actual_start_freq_hz + actual_number_freq_steps*actual_freq_step

- a) into location 8000 put the actual_start_frequencyx according to table 4 on page 6 of the manual (t4p6)
- b) into location 8000x+(4*n)x put actual_start_frequencyx +nx*actual_frequency_stepx according to t4p6, for all n's 1 to actual_number_freq_steps

b) may be accomplished in two ways. You can use the fact that the n=1'th frequency data in location 8000x+(4*(n+1))x is the data in location 8000x+(4*n)x + actual_frequency_stepx

I suggest that a table of the all the actual_frequency_n be calculated, both in hz and in binary (i.e. hex) which then can be read into the memory according to order specified in t4p6. This give one a chance to look at the table for debugging purposes. Donald thinks that data should just be programmed in on the fly.

End of Frequency Sweep Programing Description.

IQ modulation programing:

Perscription for modulation: Csech (complex sech or ln-sech) and Hermite

0) Select the modulation function and input the parameters A & a.

Csech: {A , a}={0.1 , 5}
Hermite: {A , a}={0.39714 , 2.2}

1) Input the requested dnu (in Hz) and define dw=dnu*2*pi
(dnu is the requested bandwidth (in Hz) that the modulation function will irradiate given a proper level of RF power)

2)Define dw_max = $2 \times 10^{(7)} a / (A * 512 * 2)$, dw_min= $2 \times 10^{(7)} a / (A * 2048 * 63 * 32)$
d_nu_max = dw_max/(2*pi) , d_nu_min = dw_min/(2*pi) ,
Csech: 976.563*10^3 rad/sec , 0.24420*10^3 rad/sec
155.424*10^3 Hz , 0.03855*10^3 Hz
Hermite: 108.195*10^3 rad/sec , 0.026844*10^3 rad/sec
17.220*10^3 Hz , 0.004272*10^3 Hz

3) Check that d_nu_min <= d_nu <= d_nu_max, else return an error stating that the requested d_nu in not with the available limits ... say what these limits are.

4) Compute the preliminary total number of iq points Ntiqtemp
Ntiqtemp = { a * 2x10^7 / (A * dw) } ; { ... } = nearest smaller integer,
Csech: Ntiqtemp = {10^9/dw}
Hermite: Ntiqtemp = {1.18*10^8/dw}
and confirm that 1024<= Ntiqtemp < 129024*Ncmx=4128768.

5) Assign Nc and Ncic according to the following table:

Ntiqtemp	Nc	Ncic
1024 - 2047	1	2
2048 - 4095	1	2
4096 - 8191	1	4
8192 - 16383	1	8
16384 - 32767	1	16
32768 - 65535	1	32
65536 - 129023	1	63
129024 - 258047	2	63
258048 - 516095	4	63
516096 - 1032191	8	63
1032192 - 2064383	16	63
2064384 - 4128767	32	63

Confirm that Nc <= Ncmx.

6) Compute Niq and Ntiq
Niq=[Ntiqtemp/(Nc*Ncic)] , [...] = nearest larger integer ; Ntiq = Niq*Nc*Ncic
For consistency check, 512<= Niq <= 2048, and Ntiq < 129024*Ncmx;
if they are not then the explanation/calculation below/above is inconsistent and needs to be corrected/debugged.

7) Program Niq as the argument of the n_iq function.
Program Nc and Ncic as the arguments of the Nc and icc_ir functions respectively.

8) Calculate tp :
tp = (10^-7)*Ntiq sec.

The RF on time programmed into the ppg should be = (or >) t_p .

- 9) Compute for $n=1$ to N_{iq} :
 $I(n) = \langle 511 * \text{Re}[\text{func}\{A * dw * tp\} * \{n/N_{iq} - 1/2\}] \rangle$, $\langle \dots \rangle$ means closest integer
 $Q(n) = \langle 511 * \text{Im}[\text{func}\{A * dw * tp\} * \{n/N_{iq} - 1/2\}] \rangle$

Csech: $\text{func}(x) = \text{sech}(x)^{(1+i*5)}$
 $a[n] = \text{sech}\{(dw * tp / 10) * \{n/N_{iq} - 1/2\}\}$
 $\text{phi}[n] = 5 * \ln(a(n))$
 $I[n] = \langle 511 * a(n) * \cos(\text{phi}(n)) \rangle$
 $Q[n] = \langle 511 * a(n) * \sin(\text{phi}(n)) \rangle$

Hermite: $\text{func}(x) = (1 - .957 * x^2) * \exp(-x^2)$
 $a[n] = (A * dw * tp * \{n/N_{iq} - 1/2\})^2$
 $I[n] = \langle 511 * (1 - .957 * a[n]) * \exp(-a[n]) \rangle$
 $Q[n] = 0$

- 10) Check that $I[\langle n/N_{iq}/2 \rangle]$ is either 510 or 511.
 Store the $I(n), Q(n)$ data set in decimal and 2's complement
- 11) Use the 2's complement data pairs of $(I(n), Q(n))$
 for $n=1 \rightarrow N_{iq}$ for the n_{iq} function. The same $\{I, Q\}$
 data goes in all the channels.

End of iq modulation prescription.

Explanation of iq modulation prescription:

(Indented text describes the specific case for I_n -sech case.)

i) Chose a functional shape. It can always be expressed as $f(dw * A * (t - t_p/2))$, for $0 \leq t \leq t_p$, defining the pulse width t_p , the band width dw and a scaling factor A . It is assumed that $f(0)=1$, $f(t > 0) < 1$, and $f(dw * A * t_p/2) \ll 1$. The reason for the last constraint is so that the RF power turns off properly within the defined pulse shape, otherwise power harmonics at other (non-desirable) frequencies will be introduced into the system at the end of the rf-gate.

For the I_n -sech mod function the complex modulation function is

$$w_1(t) = w_{1_max} * (\text{sech}(b * t)^{(1+i*u)}, i^2 = -1$$

or

$$w_1(t) = w_{1_max} * (\text{sech}(b * t) * \exp(i * \text{phi}(b * t)) \& \text{phi}(b * t) = u * \ln(\text{sech}(b * t))$$

The irradiated line width is $dw = 2 * u * b$ (i.e. between $\pm u * b$) and a value of $u=5$ is a good value which delivers a fairly nice selective rectangular frequency selection slice. Therefore the pulse shape is
 $f = \text{sech}(dw * .1 * (t - t_p/2))^{(1+iu)}$, for $0 \leq t \leq t_p$. i.e. $A = .1$

Let N_{iq} be the number of digitized iq pairs. The maximum N_{iq} is 2048, and we impose a minimum N_{iq} of 512 to get decent modulation shape resolution/faithfulness. Each N_{iq} pair is read (and interpolated) into the dsp in $4 * 25 * N_{cic} * N_c$ nanoseconds. Thus the entire modulation pulse width is $t_p = 10^{(-7)} * N_{iq} * N_{cic} * N_c$. Where N_c is the number of times (cycles) each iq pair is repeated as it is fed from the iq memory into the dsp modulation digitizers. The total number of 100ns points is $N_{tiq} = N_{iq} * N_{cic} * N_c$ and the form of the function in iq memory is $iq(n) = \langle 511 * f(dw * A * (n - N_{iq}/2)) \rangle$, The constant 511 reflects a 10 bit bipolar amplitude programmable in binary 2's complement format.

For the I_n -sech function the data in iq memory looks like

$$iq(n) = \langle 511 * \{\text{sech}((dw * tp / 10) * \{n/N_{iq} - .5\})\}^{(1+i*5)} \rangle,$$

$$= \langle 511 * \{\text{sech}((dw * N_{cic} * N_c * N_{iq} * 10^{(-7)} / 10) * \{n/N_{iq} - .5\})\}^{(1+i*5)} \rangle$$

To chose N_{cic} , N_c , and N_{iq} first requires relating the band width/shape of the modulation function to the pulse length. Then an approximate $N_{tiq} * t_p$ is determined so that at $n = N_{iq}$ ($t = t_p$) the function is small.

From the value of Ntiqtemp, Ncic and Nc can be determined from a table (in the previous section) and then the value of Niq (and therefore Ntiq and tp) can be determined.

ii) Ntiqtemp calculation: First define a parameter α as follows:

- a) For functions that attenuate to zero in time define α so that the value of $f(\alpha) \leq 0.015$.
- b) For functions that do not attenuate to zero define α as $N_{dnu} * A * \pi$ where N_{dnu} is the number of inverse bandwidths you require in the pulse. N_{dnu} will usually be of the order of unity in these cases.

Then Ntiqtemp is defined so that $dw * A * N_{tiqtemp} * 10^{-7} / 2 = \alpha$. i.e.

$$N_{tiqtemp} = \{ \alpha * 2 * 10^7 / (A * dw) \} = \{ 10^9 / dw \},$$

{ ... } = nearest smaller integer,

(For the ln-sech function we use $\alpha=5$, i.e. $\text{sech}(5) \approx 0.013$, which is fine.)

iii) Pick Ncic and Nc from the table. This table was produced to choose the best combination of $N_{cic} * N_c * N_{iq}$ that will deliver a faithful modulation pulse. As a guideline we tried to keep N_{iq} reasonably high to yield good resolution in the modulation line shape. However, if dw is sufficiently high, then one requires smaller $N_{cic} * N_c * N_{iq}$ to do the job. Using the guideline that the minimum acceptable N_{iq} is 512 then the product of $N_{iq} * N_c * N_{iq}$ can be categorized as in the table to cover the entire dynamic range

Ntiqtemp	Nc	Ncic
1024 - 2047	1	2
2048 - 4095	1	2
4096 - 8191	1	4
8192 - 16383	1	8
16384 - 32767	1	16
32768 - 65535	1	32
65536 - 129023	1	63
129024 - 258047	2	63
258048 - 516095	4	63
516096 - 1032191	8	63
1032192 - 2064383	16	63
2064384 - 4128767	32	63

iv) Calculate the $N_{iq} = \lceil N_{tiqtemp} / (N_c * N_{cic}) \rceil$ that is required. [...] = next largest integer.

v) Also one must ensure the frequency band requested is within the physical limits available. These limits depend on the modulation function chosen and the requirement of the smallness of f at $tp/2$.

Min $N_{tiq} = N_{tiqmn} = 2 * 512$, Max $N_{tiq} = N_{tiqmx} = 63 * 2048 * N_{cmx}$. N_{cmx} is currently 32. The relationship between N_{tiq} and dw is $dw * A * N_{tiq} * 5 * 10^{-8} = \alpha$, therefore

$$dw_{max} = 2 * 10^7 \alpha / (A * 512 * 2) \text{ (in radians/sec)},$$

$$dw_{min} = dw_{max} / (128 * N_{cmx})$$

dw must be constrained to be within this range.

for the ln-sech, $A=1$ and $\alpha = 5$ giving

$$dw_{max} = 10^9 / (512 * 2) \text{ rad/sec} = 976.56 \text{ Krad/sec} = 155.42 \text{ KHz/sec}$$

$$dw_{min} = 10^9 / (2048 * 63 * N_{cmx}) \text{ rad/sec} = 7.7505 / N_{cmx} \text{ Krad/sec} = 1.2335 / N_{cmx} \text{ KHz/sec}$$

The order of the programming will not follow the order of the explanation ..., but should follow the order of

the example implementation for the ln-sec function in the previous section.

End of iq modulation description.